

RADIO-FREQUENCY INSTRUMENT FOR THE REMOTE MONITORING OF FLOWING SOLUTIONS

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Radio-frequency (rf) devices for the concentration measurement of solutions in flow with the use of contactless probes are well known in analytical practice [1-3]. A common shortcoming of these devices is the restricted length of the cable connecting the probe in the monitored medium to the measurement and control units.

In the device proposed here the working oscillatory circuit with the probe connected into it can be set up at a considerable distance from the measurement and control units [4].

A schematic of the device is shown in Fig. 1. It includes a power supply section, a sinusoidal-voltage self-excited oscillator, the working circuit with the probe (cell), and the measurement circuitry. The supply section consists of a ferroresonance transformer Tr-1, two rectifiers comprising a bridge circuit with D-226 diode elements, and two stabilitrons D₄ and D₅ (types 817B and 915Zh). The supply section delivers the following voltages to the circuit: -20 V for the oscillator; +70 V to control the varicap for tuning of the working circuit to resonance; 5 V ac to supply the indicator light (illumination).

The sinusoidal-voltage self-excited oscillator is designed to deliver rf oscillations to the working circuit, which includes the cell. It comprises a push-pull circuit with positive feedback using P605A triode transistors T₁ and T₂. Quartz resonators K₁ and K₂ are included in the feedback loop for frequency stabilization. In the absence of the quartz resonators the frequency is stabilized by setting the oscillator to operate above a nominal frequency of 300 to 500 kHz. The resistance R₅ in the emitter circuit of the triodes T₁ and T₂ creates negative feedback and thereby reduces the nonlinear distortion. This resistance also effects temperature stabilization of the working current. The frequency of the oscillator is determined by the natural frequency of the quartz resonators, which is equal to 5 MHz in our case.

The reference voltage (detector D₁) and the voltage at the detector of the working circuit (detector D₂), which is tuned to resonance, both have the same value of 6 V. The working circuit represents a frequency-selective circuit with electronic tuning based on the use of the varicap barrier capacitance, which depends on the reverse voltage applied to it. A coax joins the output winding of the oscillator to the working circuit to supply the control voltage to the varicap.

The control voltage varies from 45 to 70 V, depending on the resonance tuning of the working circuit with the cell connected. In this case smooth tuning of the working circuit, temperature stability of the varicap, sufficient overlap with respect to the electrical parameters of the initial solutions, and the maximum possible measurement sensitivity are realized. The latter feature is attributable to the steepness of the drop at the left end of the working circuit resonance curve due to the small shift of the quasiresonance frequency.

The Q of the working circuit is increased by incomplete connection into both the circuit of the oscillator output winding and the detector circuit. For this purpose the optimum value is chosen for the coupling capacitor C₈.

The detected rf voltage is taken from the potentiometer R₉ and is transmitted to the measurement circuit. The latter represents a combination of a voltmeter with appropriate switching for tuning and subsequent preparation of the instrument for measurements of the electrical conductivity, the dielectric constant, or their sum. The poten-

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